



**Answer the following questions:**

**Q1:** (a) Describe the output voltage of the circuit shown in Fig.(1). Assuming the diodes to be actual and  $V_1 = 10 \sin \omega t$ . Sketch one cycle of the output voltage.

(b) Fig.(2) shows a simpler way to draw a transistor circuit. What are collector-emitter voltage and the transistor power dissipation?

**Q2:** The zener diode in the circuit of figure (3) has a constant reverse breakdown voltage  $V_Z = 8.2V$ , for  $75mA \leq I_Z \leq 1A$ , if  $R_L = 9 \Omega$ , size  $R_S$  so that  $V_L = V_Z$  is regulated to (maintained at)  $8.2V$  while  $V_S$  varies by  $\pm 10\%$  percent from its nominal value of  $12V$ .

**Q3:** (a) Assuming that the diodes in the circuit of Fig.(4) are ideal, find the values of the labeled voltage,  $V$ , and current,  $I$ .

(b) What is the output voltage in Fig.(5). Let  $\beta$  of the two transistors are very high.

**Q4:** The transistor in the circuit shown in fig.(6) is biased to operate in the active mode. Assuming that  $\beta$  is very large, find the collector bias current  $I_C$ . Replace the transistor with small-signal equivalent circuit T -model, find the values of the voltage gains of  $(V_{o1}/v_i)$  and  $(V_{o2}/v_i)$ .

**Q5:** For the amplifier shown in Fig.(7), let  $V_{CC} = 12V$ ,  $R_1 = 22 k\Omega$ ,  $R_2 = 6.8 k\Omega$ ,  $R_E = 560 \Omega$ , and  $R_C = 1 k\Omega$ . The transistor has  $\beta = 100$ . Calculate the dc bias current  $I_E$ . If the amplifier operates between a source for which  $R_{sig} = 600 \Omega$  and a load of  $2 k\Omega$ , replace the transistor with its hybrid- $\pi$  model, and find the values of  $R_{in}$ ,  $R_o$ , and the voltage gain  $v_o/v_{sig}$ .

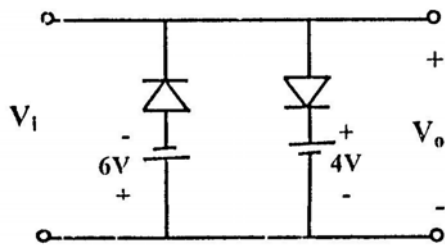


Fig. (1)

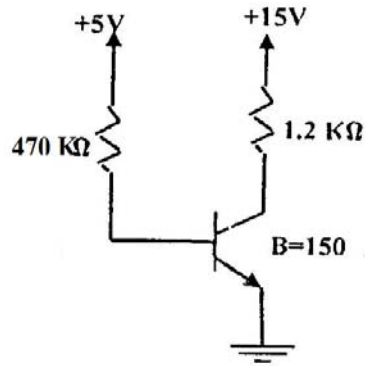


Fig.(2)

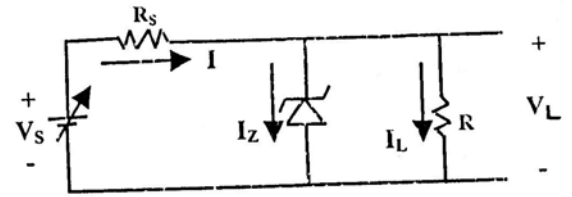


Fig.(3)

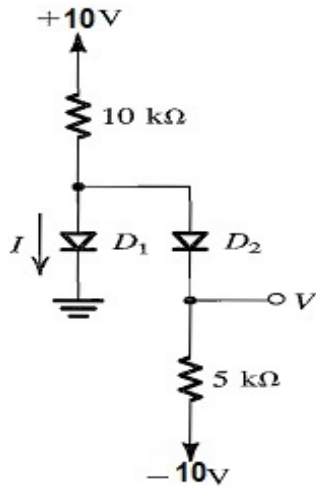


Fig.(4)

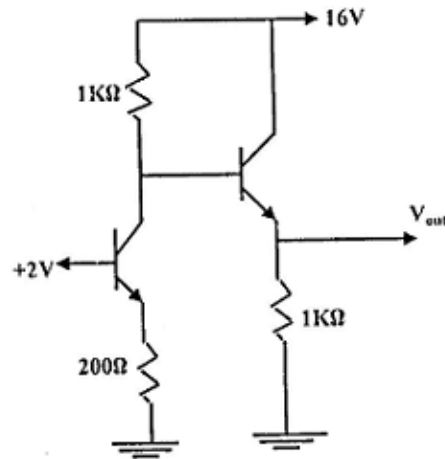


Fig.(5)

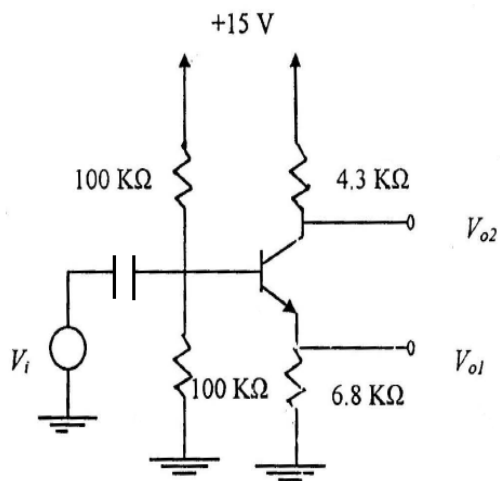


Fig.(6)

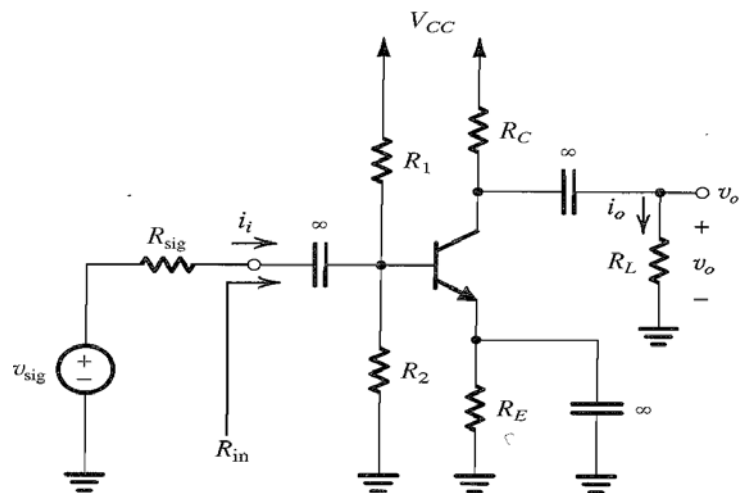


Fig.(7)

***BEST WISHES***

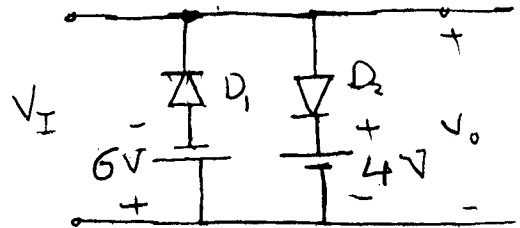
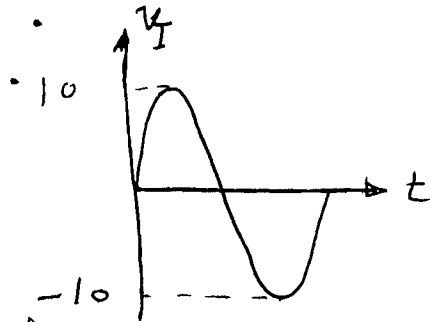
*Hossam Labib*

Q1 (a) (6 points)

Describe the o/p voltage of the circuit shown.

Assuming the diode to be actual and  $V_i = 10 \sin \omega t$ .  
Sketch one cycle of the o/p.

Solution

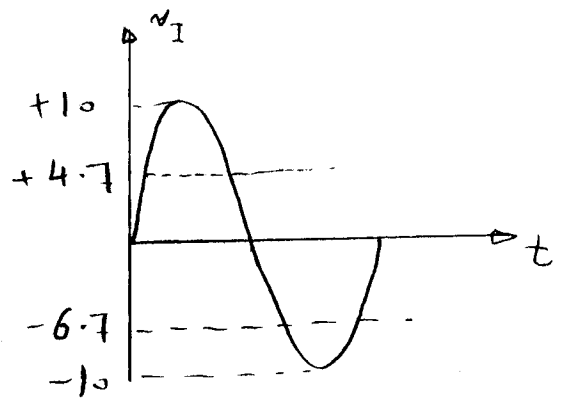
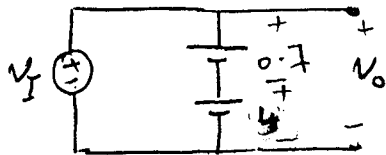


for +ve half cycle

- for  $V_i < 4.7$  V  
 $D_1$  and  $D_2$  are off  
 $\therefore V_o = V_i$

- for  $V_i > 4.7$  V (2)

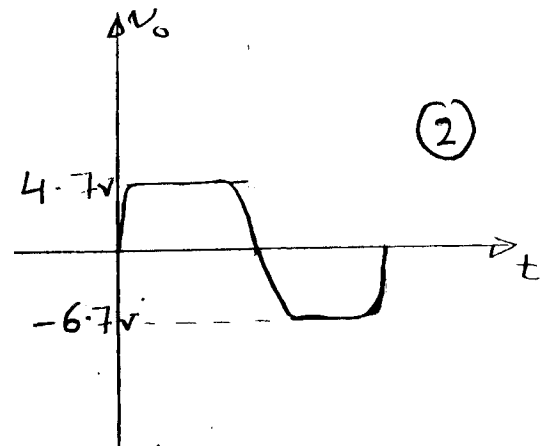
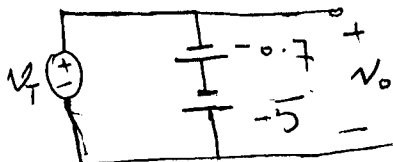
$D_1$  off &  $D_2$  on  
 $\therefore V_o = 4.7$  V



for -ve half cycle

- for  $V_i > -6.7$   
 $D_1$  off and  $D_2$  off  
 $\therefore V_o = V_i$

- for  $V_i < -6.7$  V (2)  
 $D_1$  on &  $D_2$  off  
 $\therefore V_o = -6.7$



Q<sub>1</sub>(b) (6 points)

Fig. (2) shows a simpler way to draw a Transistor circuit. What are collector emitter voltage and the Transistor power dissipation?

Solution

Let Transistor in Active Region

Loop (I)

$$5 = 470K I_B + V_{BE}$$

$$I_B = \frac{5 - 0.7}{470K} = 9.15 \mu A$$

$$\therefore I_C = \beta I_B = 150 * 9.15 \mu A = 1.3725 \text{ mA}$$

$$V_{CE} = V_{CC} - I_C R_C = 15 - 1.3725 * 1.2 = 13.353 \text{ V}$$

$$V_E = 0 \text{ V} \quad ; \quad \therefore V_{BE} = V_B - V_E$$

$$\therefore V_B = V_{BE} = 0.7 \text{ V} \quad ; \quad V_C = V_{CE} = 13.353 \text{ V}$$

$\therefore V_B > V_E \Rightarrow BE \downarrow$  Forward bias

$\therefore V_B < V_C \Rightarrow BC \downarrow$  Reverse bias

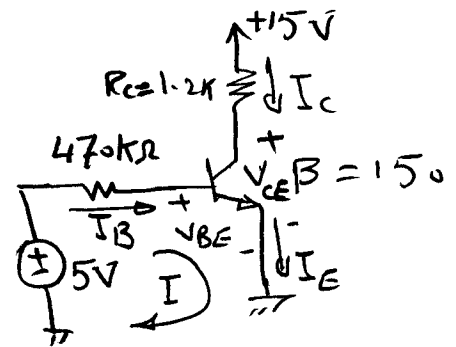
$\therefore$  Transistor in Active Region

$$\therefore V_{CE} = 13.353 \text{ V} \quad ; \quad I_C = 1.3725 \text{ mA}$$

$$P_{diss} = I_{CQ} V_{CEQ}$$

$$= 1.3725 * 10^{-3} * 13.353 = 0.0183 \text{ W}$$

$$= 18.3 \text{ mW}$$



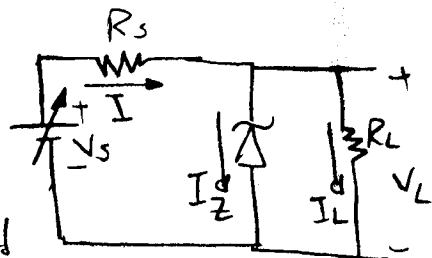
Q2: (12 points)

The Zener diode shown has a constant reverse breakdown voltage  $V_Z = 8.2 \text{ V}$

For  $75 \text{ mA} \leq I_Z \leq 1 \text{ A}$  if  $R_L = 9 \Omega$

size  $R_S$  so that  $V_L = V_Z$  is regulated

to (maintained at)  $8.2 \text{ V}$  while  $V_S$  varies by  $\pm 10\%$  percent from its nominal value of  $12 \text{ V}$



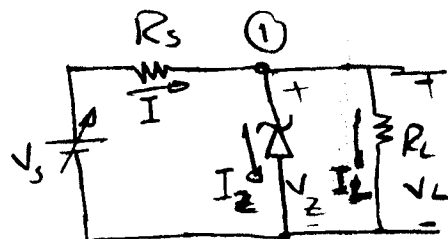
Solution

at node ①

$$I = I_Z + I_L \rightarrow \text{①}$$

$$I_L = \frac{V_L}{R_L} = \frac{V_Z}{R_L} = \frac{8.2}{9} = 0.911 \text{ A}$$

$$\therefore R_S = \frac{V_S - V_Z}{I} = \frac{V_S - V_Z}{I_Z + I_L} \rightarrow \text{②}$$



We use ② to size  $R_S$  for max. Zener current  $I_Z$  at the largest value of  $V_S$  ;  $\therefore 75 \text{ mA} \leq I_Z \leq 1 \text{ mA}$

$$\text{i.e. } I_Z = 1 \text{ A} \quad ; \quad V_S = V_S + 10\% V_S = 1.1 V_S = 1.1 \times 12 = 13.2 \text{ V}$$

$$\therefore R_S = \frac{13.2 - 8.2}{1 + 0.911} = 2.62 \Omega$$

- We check to see if  $I_Z \geq 75 \text{ mA}$  at the lowest value of  $V_S$  ; i.e.  $V_S = V_S - 10\% V_S = 0.9 V_S = 0.9 \times 12 = 10.8 \text{ V}$

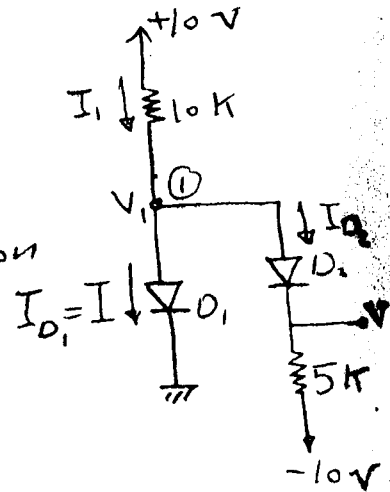
$$I_Z = I - I_L = \frac{V_S - V_Z}{R_S} - I_L = \frac{10.8 - 8.2}{2.62} - 0.911 = 0.9924 - 0.911 = \underline{81.4 \text{ mA}} > 75 \text{ mA}$$

since  $I_Z > 75 \text{ mA}$  ;  $V_Z = 8.2 \text{ V}$  then regulation is occurred (preserved)

Q3(a) (6 points)

Assuming that the diodes in the circuit of Fig.(4) are ideal, Find the values of the labeled voltage,  $V$ , and current,  $I$ .

Solution



let  $I = I_{D1}$

let  $D_1$  and  $D_2$  are on

For  $D_1$  on  $\Rightarrow V_1 = 0 \Rightarrow V = 0$  for  $D_2$  on

$$\therefore I_1 = \frac{10 - V_1}{10k} = \frac{10 - 0}{10k} = 1mA$$

$$I_{D2} = \frac{V_1 - (-10)}{5k} = \frac{0 + 10}{5k} = 2mA$$

From Node ①  $\Rightarrow I_1 = I_{D1} + I_{D2}$

$$I_{D1} = I_1 - I_{D2} = 1mA - 2mA = -1mA$$

$\therefore I_{D1}$  -ve  $\therefore$  NOT True

i.e. the assumption NOT correct

Let  $D_1$  off and  $D_2$  on

- For  $D_1$  off  $\Rightarrow I_{D1} = 0$  ;  $I_1 = I_{D2}$

$$\text{- For } D_2 \text{ on } \Rightarrow I_1 = \frac{10 - (-10)}{10k + 5k} = \frac{20}{15k} = 1.333mA$$

$$V = I_{D2} * 5k - 10 = 1.333 * 5 - 10 = -3.335V$$

For  $V = -3.335V$  ;  $\therefore V_1 = V = V_{p1} = -3.335$  ;  $V_{n1} = 0$

i.e.  $V_{p1} < V_{n1} \Rightarrow D_1$  off

$\therefore$  The assumption is True ( $D_1$  off &  $D_2$  on)

$$\therefore V = -3.335V \text{ and } I = I_{D1} = 0A$$

Q3(b): (6 points)

What is the o/p voltage.

- Let  $\beta$  of the 2-transistors is very large.

Solution

$\therefore \beta$  is very high

$$\therefore I_{B1} = I_{B2} = 0$$

Loop (I) let  $Q_1$  and  $Q_2$  in Active region

$$-2 + V_{BE1} + I_{E1} R_{E1} = 0$$

$$I_{E1} = \frac{2 - 0.7}{200} = 6.5 \text{ mA}$$

$$\therefore I_{C1} = I_{E1}$$

$$\therefore \text{at node } \textcircled{1} \Rightarrow I_1 = I_{C1} + I_{B1} \quad ; \quad I_{B2} = 0$$

$$\therefore I_1 = I_{C1} = 6.5 \text{ mA}$$

$$V_{C1} = 16 - I_1 \times 1K = 16 - 6.5 \times 1 = 9.5 \text{ V} = V_{B2}$$

$$V_{E1} = I_{E1} R_{E1} = 6.5 \times 0.2 = 1.3 \text{ V}$$

Loop (II)

$$-V_{C1} + V_{BE2} + I_{E2} R_{E2} = 0$$

$$I_{E2} = \frac{9.5 - 0.7}{1K} = 8.8 \text{ mA}$$

$$\therefore V_0 = I_{E2} R_{E2} = 8.8 \times 1 = 8.8 \text{ V} = V_{E2}$$

$$V_{C2} = 16 \text{ V}$$

$$\text{for } Q_1 \quad \therefore V_{B1} > V_{E1} \Rightarrow \text{BES} \rightarrow \text{Forward}$$

$$V_{B1} < V_{C1} \Rightarrow \text{BCJ} \rightarrow \text{Reverse}$$

$\therefore Q_1$  in Active region

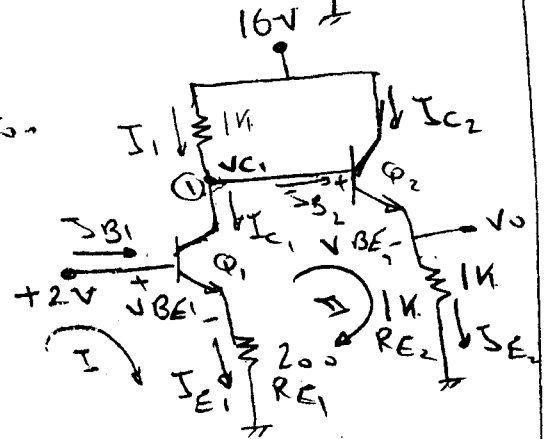
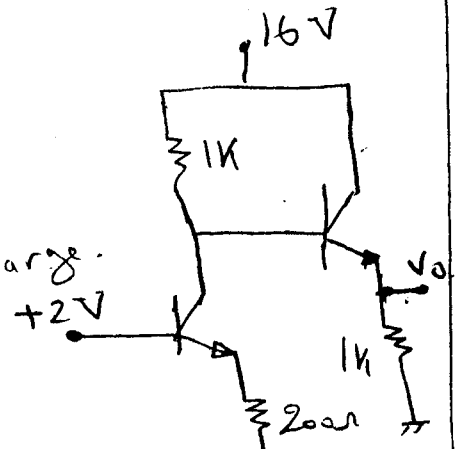
$$\text{for } Q_2 \quad \therefore V_{B2} > V_{E2} \Rightarrow \text{BES} \rightarrow \text{Forward}$$

$$V_{B2} < V_{C2} \Rightarrow \text{BCJ} \rightarrow \text{Reverse}$$

$\therefore Q_2$  in Active region

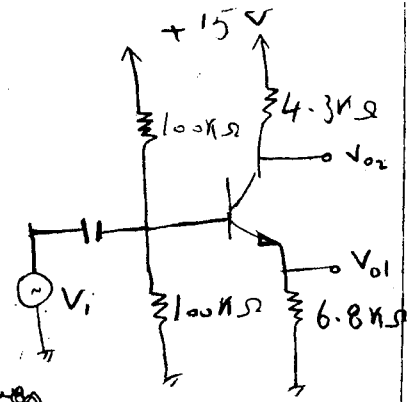
$\therefore$  Assumption is True.

$$\therefore V_0 = V_{E2} = 8.8 \text{ V}$$



Q4: (12 points)

The transistor in the circuit shown is biased to operate in the active mode. Assuming that  $\beta$  is very large, find the collector bias current. Replace the transistor with small-signal equivalent circuit T-Model, find the values of the voltage gains of  $V_{o1}$ ,  $V_{o2}$ .



DC analysis

Reduce AC sources i.e  $V_i = s.c$   
and ALL capacitor are o.c

$$V_{Th} = 15 \frac{100k}{100k + 100k} = 7.5 \text{ V}$$

$$R_{Th} = 100k \parallel 100k = 50 \text{ k}\Omega$$

$\because \beta$  is very large  $\Rightarrow I_B = 0$

$$\alpha = 1 \Rightarrow I_C = I_E$$

Loop (I)

$$-V_{Th} + I_B R_{Th} + V_{BE} + 6.8k I_E = 0$$

$$I_E = \frac{V_{Th} - V_{BE}}{6.8k} = \frac{7.5 - 0.7}{6.8k} = 1 \text{ mA}$$

$$\therefore I_C = 1 \text{ mA} \quad \text{--- (2)}$$

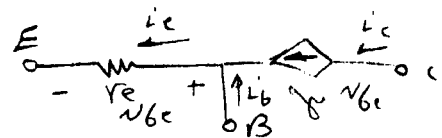
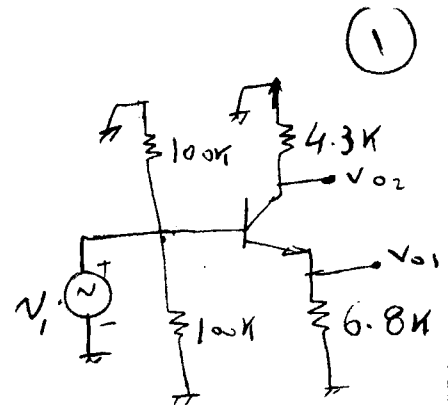
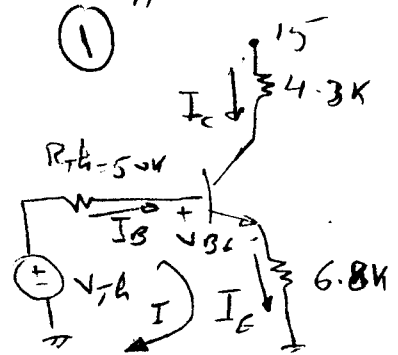
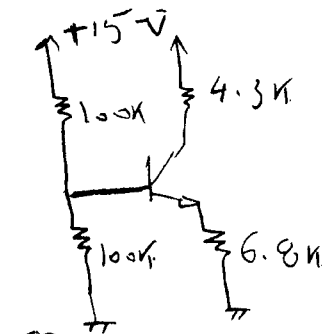
AC analysis

Reduce DC sources  $V_{CC} = s.c$   
ALL capacitor are s.c

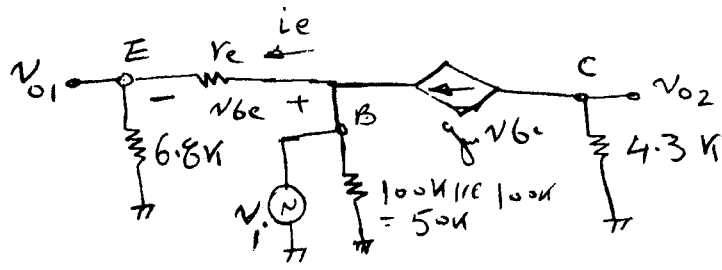
The parameters of T-Model are

$$g_m = \frac{I_C}{V_T} = \frac{1 \text{ mA}}{25 \text{ mV}} = 40 \text{ mA/V}$$

$$r_e = \frac{\alpha}{g_m} \approx \frac{1}{40} = 25 \Omega$$







$$V_{02} = -g_m V_{be} \times 4.3K \rightarrow \textcircled{2}$$

$$V_{be} = V_i \cdot \frac{r_e}{r_e + 6.8K} \rightarrow \textcircled{1}$$

$$V_{02} = -g_m V_i \cdot \frac{r_e \times 4.3K}{r_e + 6.8K}$$

$$\frac{V_{02}}{V_i} = -g_m \cdot \frac{r_e \times 4.3K}{r_e + 6.8K} = -40 \times 10^3 \cdot \frac{25 \times 4300}{25 + 6800} = -0.63 \text{ V/V}$$

$$\# V_{01} = V_i \cdot \frac{6.8K}{6.8K + r_e} \rightarrow \textcircled{2}$$

$$\frac{V_{01}}{V_i} = \frac{6.8K}{6.8K + 25} = 0.996 \text{ V/V} \textcircled{1}$$

**Q5: (12 points)**

For the amplifier shown in Fig.(3), let  $V_{CC} = 12V$ ,  $R_1 = 22\text{ k}\Omega$ ,  $R_2 = 6.8\text{ k}\Omega$ ,  $R_E = 560\ \Omega$ , and  $R_C = 1\text{ k}\Omega$ . The transistor has  $\beta = 100$ . Calculate the dc bias current  $I_E$ . If the amplifier operates between a source for which  $R_{sig} = 600\ \Omega$  and a load of  $2\text{ k}\Omega$ , replace the transistor with its hybrid- $\pi$  model, and find the values of  $R_{in}$ ,  $R_o$ , and the voltage gain  $v_o/v_{sig}$ .

Solution

Dc Analysis

- ALL capacitor are o-c
- Reduce AC sources

$$R_{Th} = R_1 \parallel R_2 = 22\text{K} \parallel 6.8\text{K}$$

$$R_{Th} = 5.19\text{K}\ \Omega$$

$$V_{Th} = V_{CC} \frac{R_2}{R_1 + R_2}$$

$$V_{Th} = 12 \frac{6.8\text{K}}{22\text{K} + 6.8\text{K}} = 2.833\text{ V}$$

Loop (I)

$$-V_{Th} + I_B R_{Th} + V_{BE} + I_E R_E = 0$$

$$\therefore I_E = (1 + \beta) I_B$$

$$\frac{I_E}{1 + \beta} R_{Th} + I_E R_E = V_{Th} - V_{BE}$$

$$I_E = \frac{V_{Th} - V_{BE}}{R_E + \frac{R_{Th}}{1 + \beta}} = \frac{2.833 - 0.7}{560 + \frac{5.19\text{K}}{101}} = 3.489\text{ mA} \quad \text{--- (1)}$$

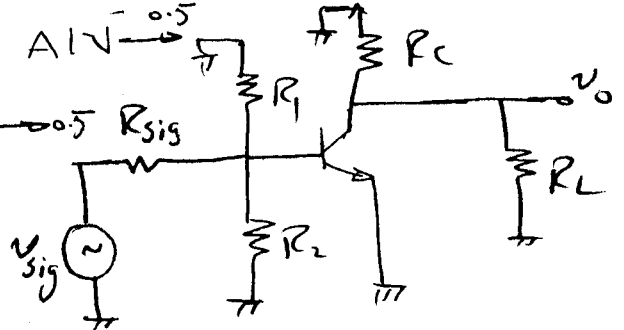
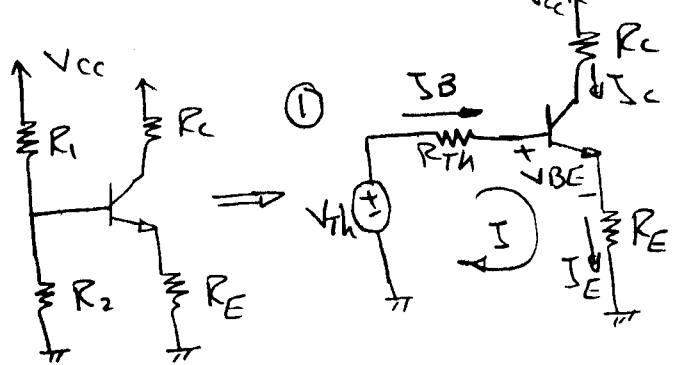
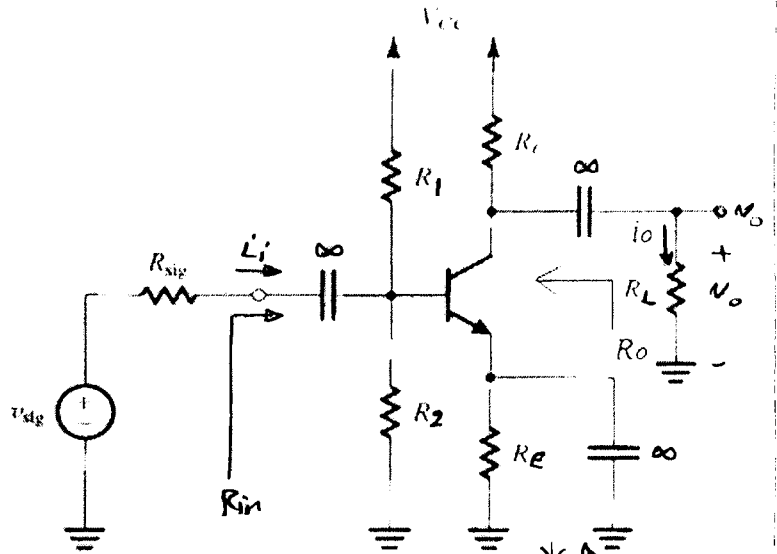
$$I_C = \alpha I_E = \frac{\beta}{1 + \beta} I_E = \frac{100}{101} \times 3.489\text{ mA} = 3.454\text{ mA}$$

$$g_m = \frac{I_C}{V_T} = \frac{3.454\text{ mA}}{25\text{ mV}} = 0.138\text{ A/V}$$

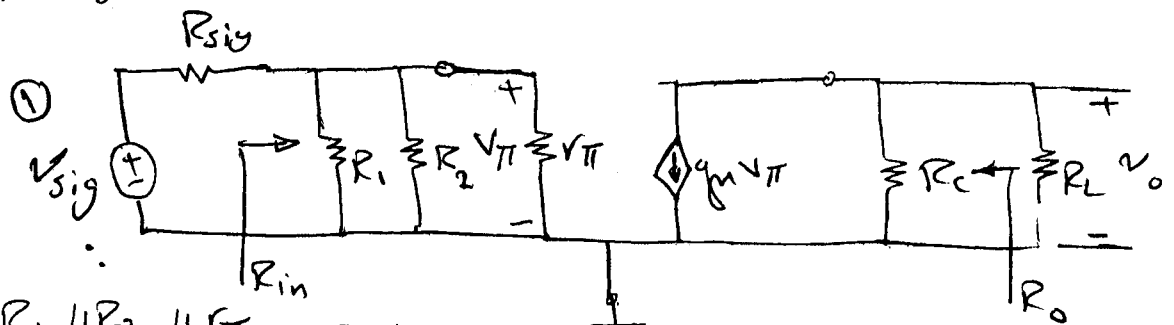
$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{0.138} = 724.64\ \Omega$$

AC Analysis

- ALL capacitor are s.c
- Reduce DC sources



Using  $\pi$ -model



$$R_{in} = R_1 \parallel R_2 \parallel r_{\pi} = 22K \parallel 6.8K \parallel 724.61$$

$$= 635.93 \approx 636 \Omega \quad \longrightarrow \textcircled{2}$$

$$R_{o|_{v_{sig}=0}} = R_C = 1 K\Omega. \quad \longrightarrow \textcircled{2}$$

To find  $\frac{v_o}{v_{sig}}$  :-

$$v_o = -g_m v_{\pi} (R_C \parallel R_L)$$

$$v_{\pi} = v_{sig} \frac{R_{in}}{R_{in} + R_{sig}}$$

$$v_o = -g_m v_{sig} \frac{R_{in}}{R_{in} + R_{sig}} (R_C \parallel R_L) \quad \longrightarrow \textcircled{2}$$

$$\frac{v_o}{v_{sig}} = -g_m \frac{R_{in}}{R_{in} + R_{sig}} (R_C \parallel R_L)$$

$$= -0.138 \frac{636}{636 + 600} (1K \parallel 2K)$$

$$= -0.138 * 0.515 * 666.667 = -47.34 \text{ V/V} \quad \longrightarrow \textcircled{1}$$